
Progressive Algebraic Soft-Decision Decoding of Reed-Solomon Codes

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Outline



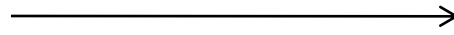
- Introduction
- Progressive ASD (PASD) algorithm
- Validity analysis
- Complexity analysis
- Error-correction performance
- Conclusions

I. Introduction of list decoding

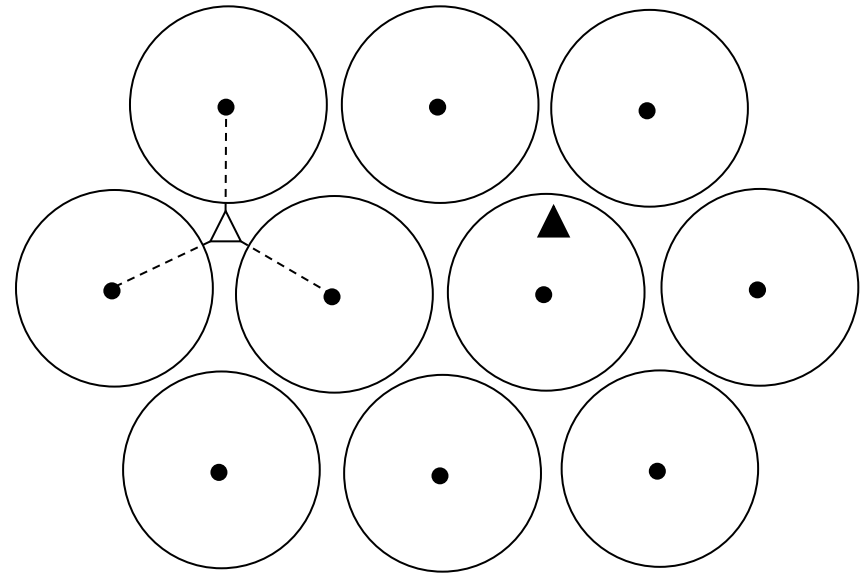
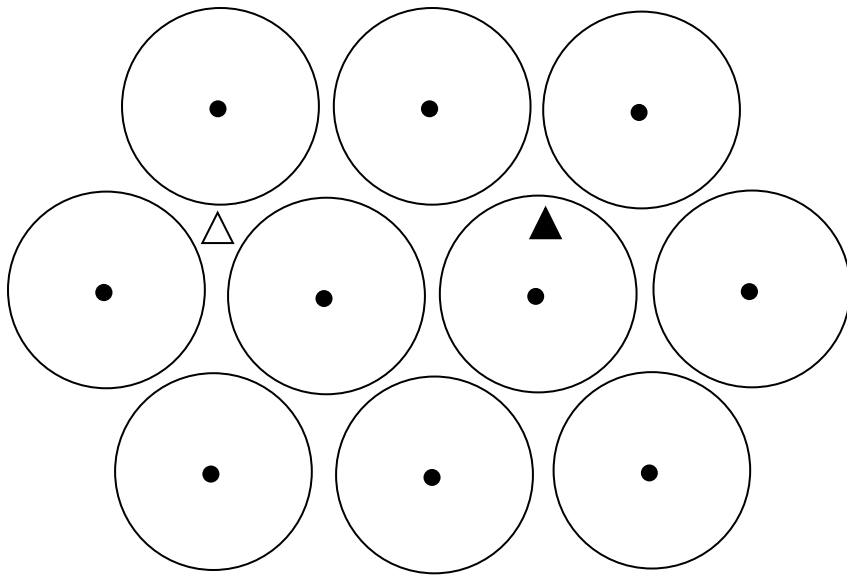


- Decoding philosophy evolution of an (n, k) RS code

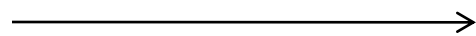
Unique decoding



List decoding



$$\tau_{\text{unique}} = \left\lfloor \frac{n-k}{2} \right\rfloor$$



$$\tau_{\text{list}} = n - \left\lfloor \sqrt{n(k-1)} \right\rfloor - 1$$

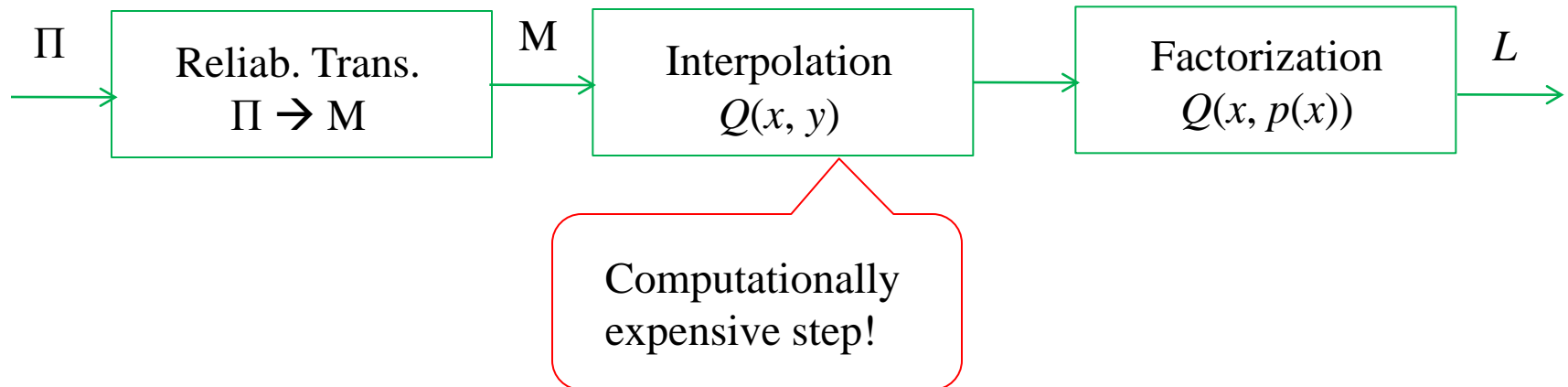
I. Overview of Enc. & Decd.



■ Encoding

- Given a message polynomial: $u(x) = u_0 + u_1x + \dots + u_{k-1}x^{k-1}$ ($u_i \in \text{EGF}(q)$)
- Generate the codeword of an (n, k) RS code
$$\mathbf{c} = (c_0, c_1, \dots, c_{n-1}) = (u(\alpha_0), u(\alpha_1), \dots, u(\alpha_{n-1}))$$
 ($\alpha_i \in \text{EGF}(q) \setminus \{0\}$)

■ Decoding

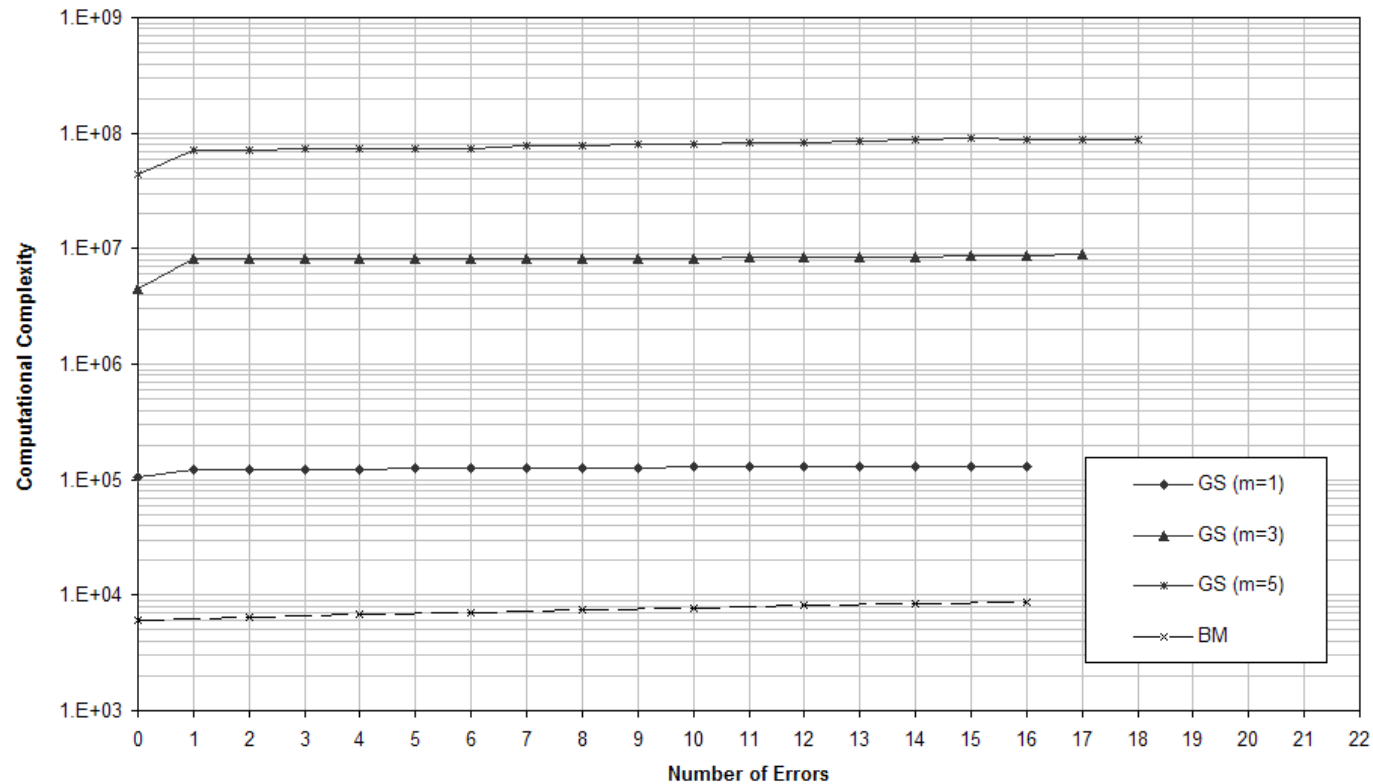


I. Introduction



- Price to pay: decoding complexity

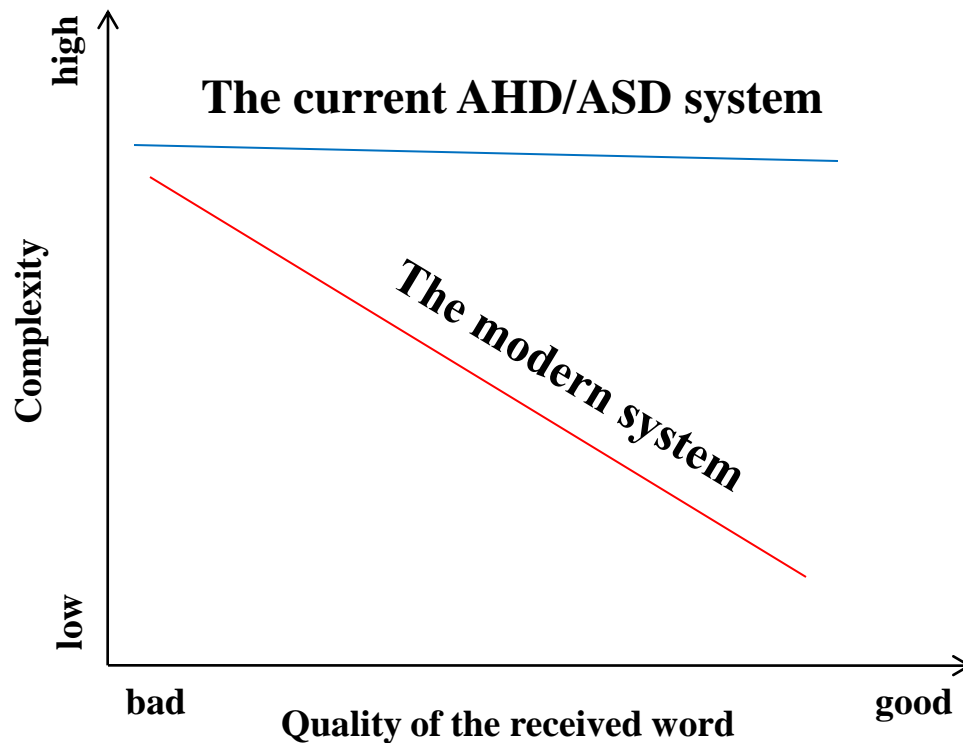
Decoding complexity of GS decoding of RS (63, 31) code



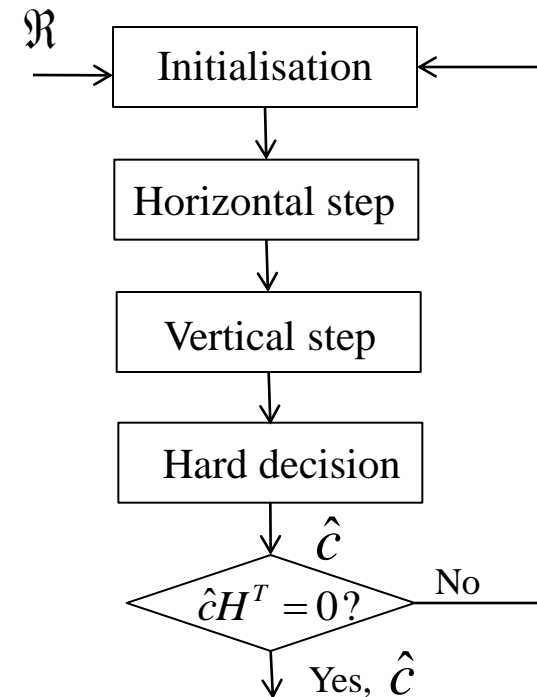
I. Inspirations



- The algebraic soft decoding is of high complexity. It is mainly due to the iterative *interpolation process*;
- A modernized thinking – the decoding should be *flexible* (i.e., channel dependent).



E.g., the belief propagation algorithm



Iterative proc.



Incremental comp.

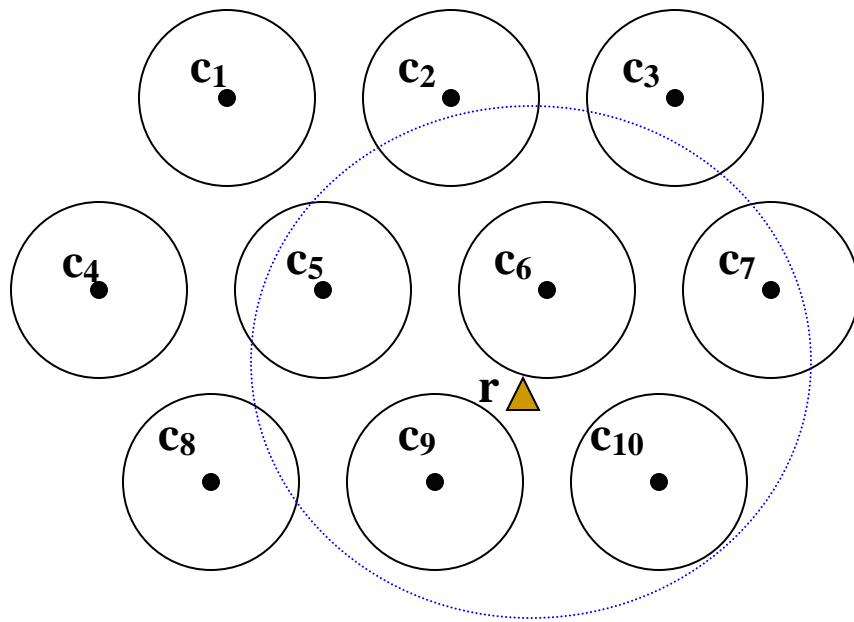


Continues validat.

II. A Graphical Introduction

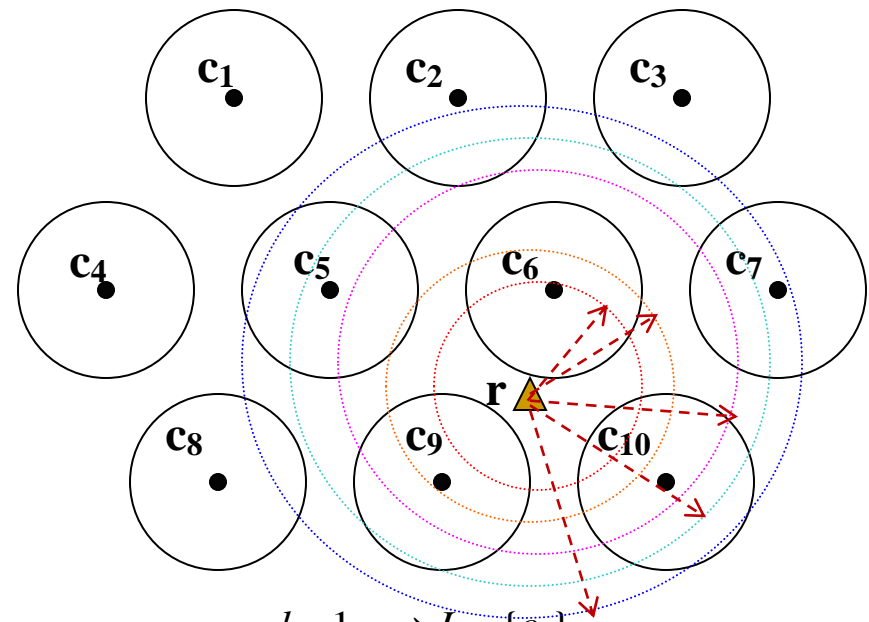


ASD ($l=5$) \longrightarrow PASD ($l=1 \rightarrow 2 \rightarrow 3 \rightarrow 4 \rightarrow 5$)



$$L = \{c_5, c_6, c_7, c_9, c_{10}\}$$

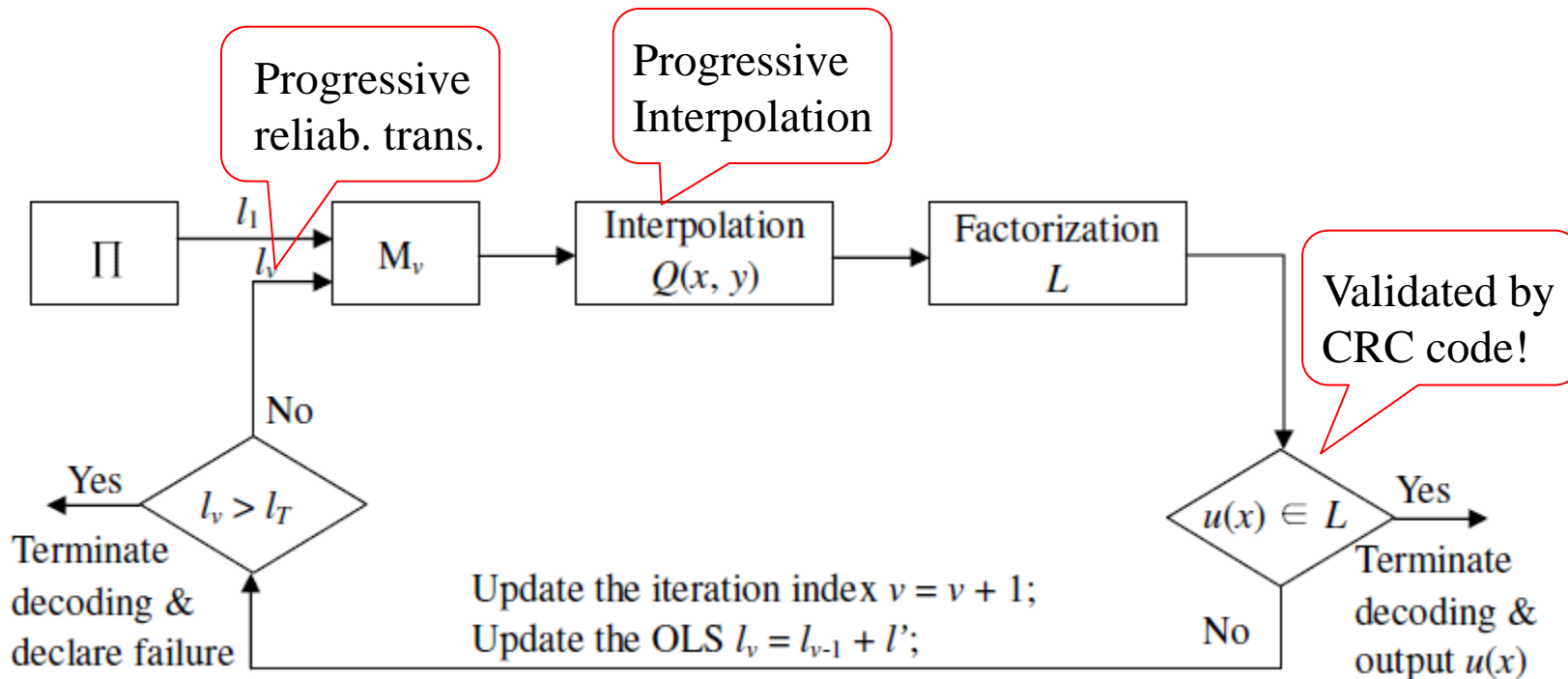
$|L|$ - factorization output list size



- $l=1 \rightarrow L = \{c_6\}$
- $l=2 \rightarrow L = \{c_6, c_9\}$
- $l=3 \rightarrow L = \{c_6, c_9, c_{10}\}$
- $l=4 \rightarrow L = \{c_5, c_6, c_9, c_{10}\}$
- $l=5 \rightarrow L = \{c_5, c_6, c_7, c_9, c_{10}\}$

Enlarging the decoding radius progressively \rightarrow Enlarging the factorization OLS progressively

II. Decoding architecture



v – iteration index;

l_v -- designed OLS at each iteration;

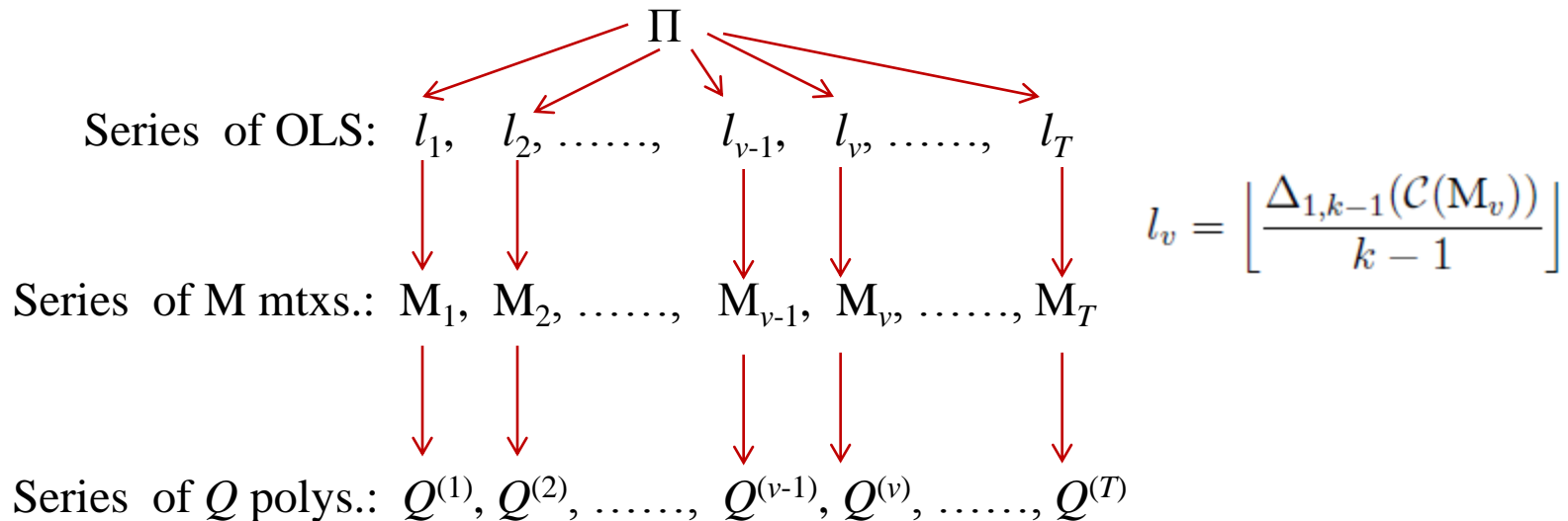
l_T -- designed maximal OLS (~the maximal complexity that the system can tolerate);

l' – step size for updating the OLS, $l_{v+1} = l_v + l'$;

II. Progressive approach



- Enlarge the decoding radius \rightarrow Enlarge the OLS
- Progressive decoding



Question: Can the solution of $Q^{(v)}$ be found based on the knowledge of $Q^{(v-1)}$?

II. Incremental interpolation constraints

- Multiplicity $m_{ij} \sim$ interpolated point (x_j, α_i)
- Given a polynomial $Q(x, y)$, m_{ij} implies $\frac{1}{2}m_{ij}(m_{ij} + 1)$ constraints of

$$\mathcal{D}_{r,s}(Q(x, y))|_{x=x_j, y=\alpha_i} = \sum_{a \geq r, b \geq s} \binom{a}{r} \binom{b}{s} Q_{ab} x_j^{a-r} \alpha_i^{b-s} = 0 \quad (r + s < m_{ij})$$

- **Definition 1:** Let $\Lambda(m_{ij})$ denote a set of interpolation constraints $(r, s)_{ij}$ indicated by m_{ij} , then $\Lambda(\mathbf{M})$ denotes a collection of all the sets $\Lambda(m_{ij})$ defined by the entry m_{ij} of \mathbf{M}

$$\Lambda(\mathbf{M}) = \{\Lambda(m_{ij}), \forall m_{ij} \in \mathbf{M}\}$$

- Example: $\mathbf{M} = \begin{bmatrix} 2 & 0 & 0 \\ 0 & 1 & 1 \\ 1 & 2 & 0 \\ 0 & 0 & 1 \end{bmatrix} \implies \Lambda(\mathbf{M}) = \{(0, 0)_{00}, (1, 0)_{00}, (0, 1)_{00}, (0, 0)_{11}, (0, 0)_{12}, (0, 0)_{20}, (0, 0)_{21}, (1, 0)_{21}, (0, 1)_{21}, (0, 0)_{32}\}$

II. Incremental Interp. Constr.



- Definition 2:** Let m_{ij}^{v-1} and m_{ij}^v denote the entries of matrix M_{v-1} and M_v , the incremental interpolation constraints introduced between the matrices are defined as a collection of all the residual sets between $\Lambda(m_{ij}^v)$ and $\Lambda(m_{ij}^{v-1})$ as:

$$\Lambda(\Delta M_v) = \{\Lambda(M_v) \setminus \Lambda(M_{v-1})\} = \{\Lambda(m_{ij}^v) \setminus \Lambda(m_{ij}^{v-1}), \forall m_{ij}^v \in M_v \text{ and } m_{ij}^{v-1} \in M_{v-1}\}$$

- Example:**

$$M_2 = \begin{bmatrix} 2 & 0 & 0 \\ 0 & 1 & 1 \\ 1 & 2 & 0 \\ 0 & 0 & 1 \end{bmatrix} \qquad M_1 = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 0 & 1 \\ 1 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$\Lambda(M_2) = \{(0, 0)_{00}, (1, 0)_{00}, (0, 1)_{00}, (0, 0)_{11}, (0, 0)_{12}, (0, 0)_{20}, (0, 0)_{21}, (1, 0)_{21}, (0, 1)_{21}, (0, 0)_{32}\}$
 $\Lambda(M_1) = \{(0, 0)_{00}, (0, 0)_{12}, (0, 0)_{20}, (0, 0)_{21}, (0, 0)_{32}\}$

$\Lambda(\Delta M_2) = \{(1, 0)_{00}, (0, 1)_{00}, (0, 0)_{11}, (1, 0)_{21}, (0, 1)_{21}\}$

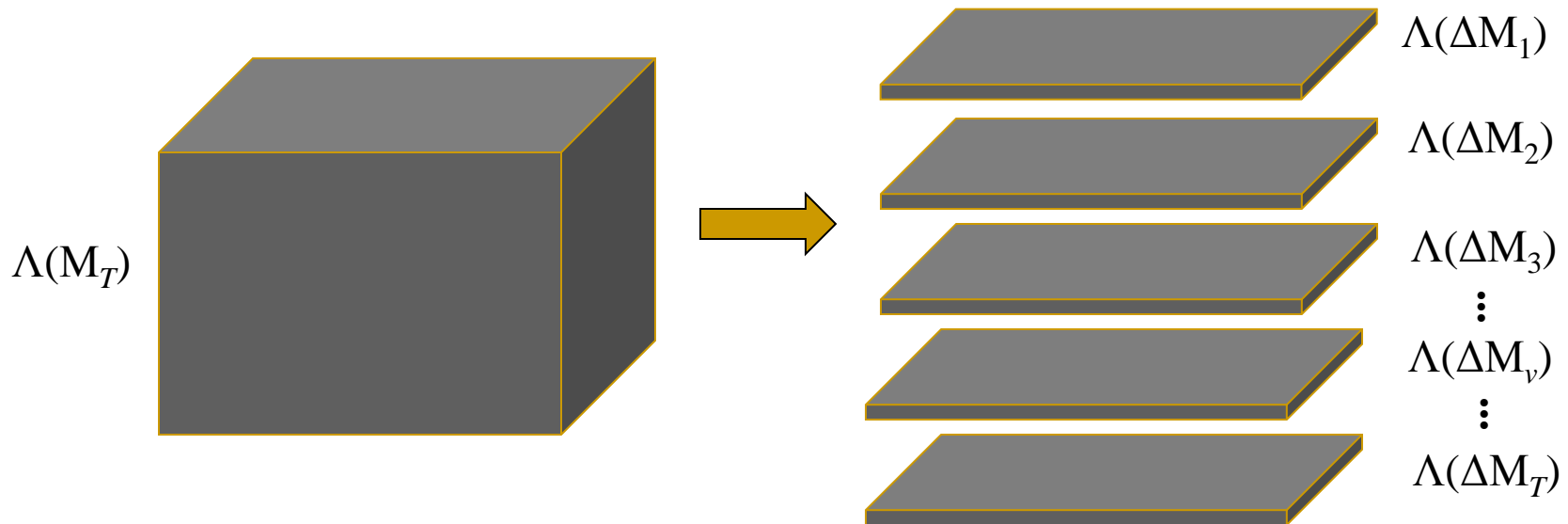
5 constraints
5 constraints

10 constraints

II. Progressive Interpolation



- A big chunk of interpolation task $\Lambda(M_T)$ can be sliced into smaller pieces.



- $\Lambda(M_T) = \Lambda(\Delta M_1) \cup \Lambda(\Delta M_2) \cup \dots \cup \Lambda(\Delta M_v) \cup \dots \cup \Lambda(\Delta M_T)$.
- $\Lambda(\Delta M_v)$ defines the interpolation task of iteration v .

II. Incremental Computations



- **Review** on the interpolation process – iterative polynomial construction
- Given M_v , the interpolation constraints are $\Lambda(M_v)$
- The polynomial group is: $\mathbf{G}_v = \{g_0, g_1, \dots, g_v\}$, ($\deg_y g_v = l_v$)
- The iterative process

For $(r, s)_{ij} \in \Lambda(M_v)$

$$f_{(r,s)_{ij}} = \min\{g_t \mid \mathcal{D}_{(r,s)_{ij}}(g_t) \neq 0\}.$$

$f_{(r,s)_{ij}}$ defines the outcome of the updated group.

For each $g_t \in \mathbf{G}_v$

$$g_t = \begin{cases} g_t, & \text{if } \mathcal{D}_{(r,s)_{ij}}(g_t) = 0 \\ [g_t, f_{(r,s)_{ij}}]_D, & \text{if } \mathcal{D}_{(r,s)_{ij}}(g_t) \neq 0 \text{ and } g_t \neq f_{(r,s)_{ij}} \\ [x f_{(r,s)_{ij}}, f_{(r,s)_{ij}}]_D, & \text{if } f_{(r,s)_{ij}}, \end{cases}$$

Finally, $Q^{(v)} = \min\{g_t \mid g_t \in \mathbf{G}_v\}$

II. Incremental Computations



- From iteration $\nu-1$ to $\nu \dots$
- The progressive interpolation can be seen as a ***progressive polynomial group expansion*** which consists of two successive stages.
- Let $\tilde{\mathbf{G}}_{\nu-1} = \{\tilde{g}_0, \tilde{g}_1, \dots, \tilde{g}_{l_{\nu-1}}\}$ be the outcome of iteration $\nu-1$.
- During the generation of $\tilde{\mathbf{G}}_{\nu-1}$, a series of $f_{(r,s)_{ij}}$ with $(r, s)_{ij} \in \Lambda(\mathbf{M}_{\nu-1})$ are identified and stored.
- ***Expansion I:*** expand the ***number*** of polynomials of the group

$$\Delta\mathbf{G}_\nu = \{g_{l_{\nu-1}+1}, \dots, g_{l_\nu}\} = \{y^{l_{\nu-1}+1}, \dots, y^{l_\nu}\}$$

- Polynomials of $\Delta\mathbf{G}_\nu$ perform interpolation w.r.t. constraints of $\Lambda(\mathbf{M}_{\nu-1})$;
- Polynomials $f_{(r,s)_{ij}}$ are re-used for the update of $\Delta\mathbf{G}_\nu$.
- Let $\Delta\tilde{\mathbf{G}}_\nu$ be the updated outcome of $\Delta\mathbf{G}_\nu$ and $\mathbf{G}_\nu = \tilde{\mathbf{G}}_{\nu-1} \cup \Delta\tilde{\mathbf{G}}_\nu$.

II. Incremental Computations

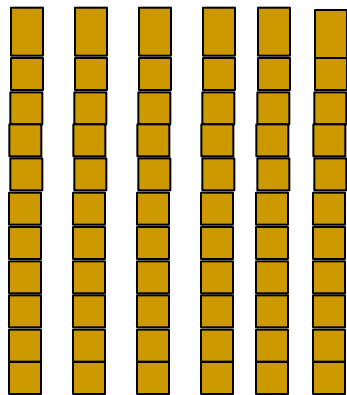


- **Expansion II:** expand the *size* of polynomials of the group \mathbf{G}_v
- Polynomials of \mathbf{G}_v will now perform interpolation w.r.t. the incremental constraints $\Lambda(\Delta M_v)$, yielding $\tilde{\mathbf{G}}_v$.
- Finally, $Q^{(v)}(x, y) = \min\{\tilde{g}_t \mid \tilde{g}_t \in \tilde{\mathbf{G}}_v\}$

- → Visualize the polynomial group expansion *Expansion I*

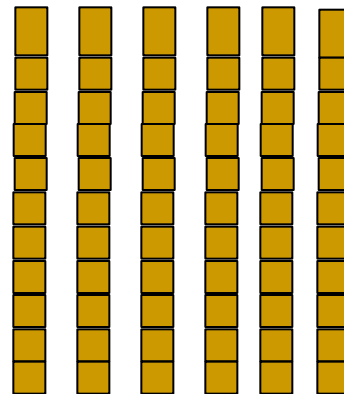
E.g., $l_v = 5$

g_1 g_2 g_3 g_4 g_5 g_6



ASD

g_1 g_2 g_3 g_4 g_5 g_6



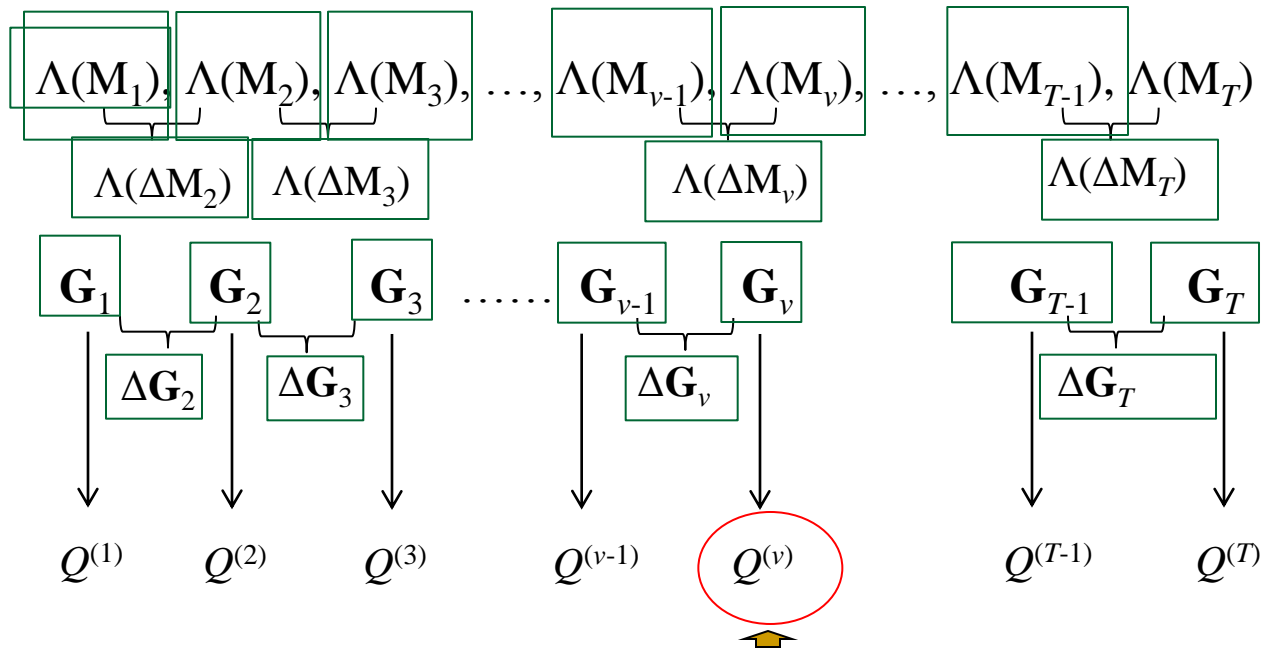
PASD

Expansion II

II. Progressive Interpolation



- The process of progressive interpolation
- $M_1, M_2, M_3, \dots, M_{v-1}, M_v, \dots, M_{T-1}, M_T$



If $Q^{(v)}(x, u(x)) = 0$, the decoding will be terminated.

- Multiple factorizations are carried out in order to determine whether $u(x)$ has been found!

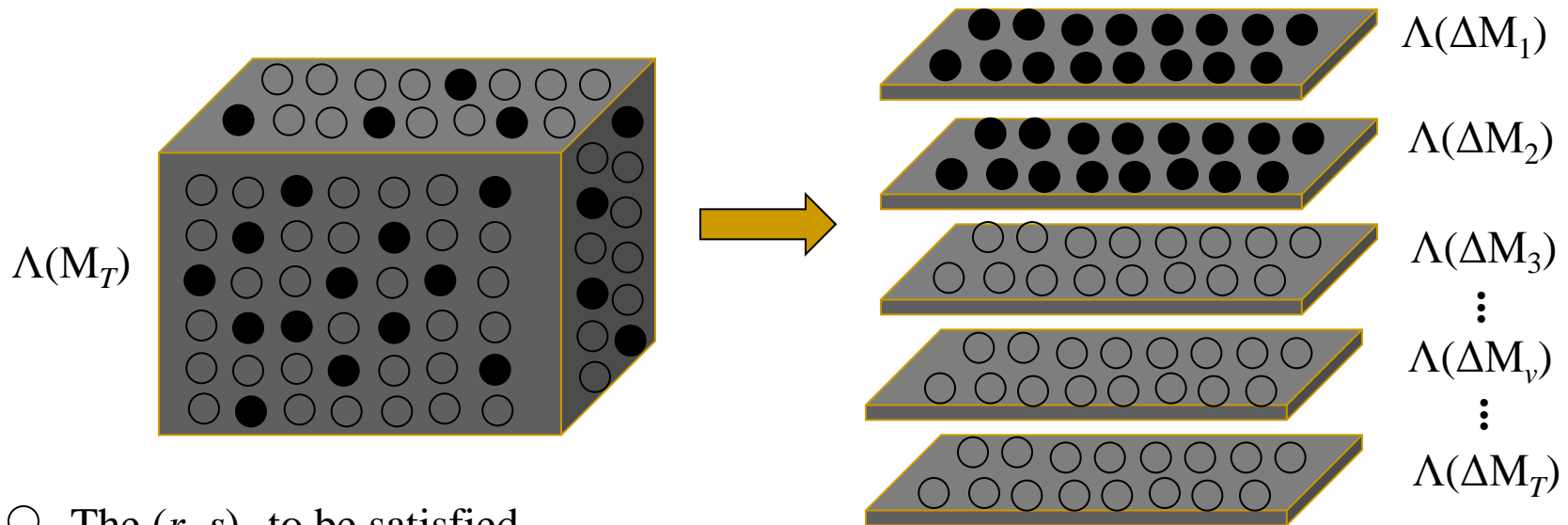
III. Validity Analysis



- For any two $(r_1, s_1)_{i_1j_1}$ and $(r_2, s_2)_{i_2j_2}$ of $\Lambda(M_T)$:

$$(r_1, s_1)_{i_1j_1} \longrightarrow (r_2, s_2)_{i_2j_2} \iff (r_2, s_2)_{i_2j_2} \longrightarrow (r_1, s_1)_{i_1j_1}$$

- The algorithm imposes a progressive interpolation order



- The $(r, s)_{ij}$ to be satisfied.
- The satisfied $(r, s)_{ij}$.

III. Validity Analysis



- Decoding with an OLS of l_v , the solution $Q(x, y)$ is seen as the minimal candidate chosen from the cumulative kernel $\overline{\mathcal{K}_{\mathcal{C}(M_v)}}$

$$\overline{\mathcal{K}_{\mathcal{C}(M_v)}} = \{Q \in \mathbb{F}_q[x, y] \mid \mathcal{D}_w(Q) = 0 \forall w \in \Lambda(M_v), \deg_y Q \leq l_v\}$$

$$Q(x, y) = \min\{Q \in \overline{\mathcal{K}_{\mathcal{C}(M_v)}}\}$$

- For both of the algorithms:

$$\Lambda(M_v) = \Lambda(\Delta M_1) \cup \Lambda(\Delta M_2) \cup \dots \cup \Lambda(\Delta M_v)$$

the same set of constraints are defined for the cumulative kernel;

- Consequently, they will offer the same solution of $Q(x, y)$.

III. Validity Analysis



- In the end of *Expansion I*, $\mathbf{G}_v = \tilde{\mathbf{G}}_{v-1} \cup \Delta\tilde{\mathbf{G}}_v$
- Can $\tilde{\mathbf{G}}_{v-1}$ and $\Delta\tilde{\mathbf{G}}_v$ be found separately?
- Recall the polynomial updating rules

$$g_t = \begin{cases} g_t, & \text{if } \mathcal{D}_{(r,s)ij}(g_t) = 0 \\ [g_t, f_{(r,s)ij}]_D, & \text{if } \mathcal{D}_{(r,s)ij}(g_t) \neq 0 \text{ and } g_t \neq f_{(r,s)ij} \\ [x f_{(r,s)ij}, f_{(r,s)ij}]_D, & \text{if } f_{(r,s)ij}, \end{cases}$$

- The minimal polynomial $f_{(r,s)ij}$ defines the solution of one round of poly. update w.r.t. $(r, s)_{ij}$.
- If such a group expansion procedure does not change the identity of $f_{(r,s)ij}$, $\tilde{\mathbf{G}}_{v-1}$ and $\Delta\tilde{\mathbf{G}}_v$ can indeed be found separately.

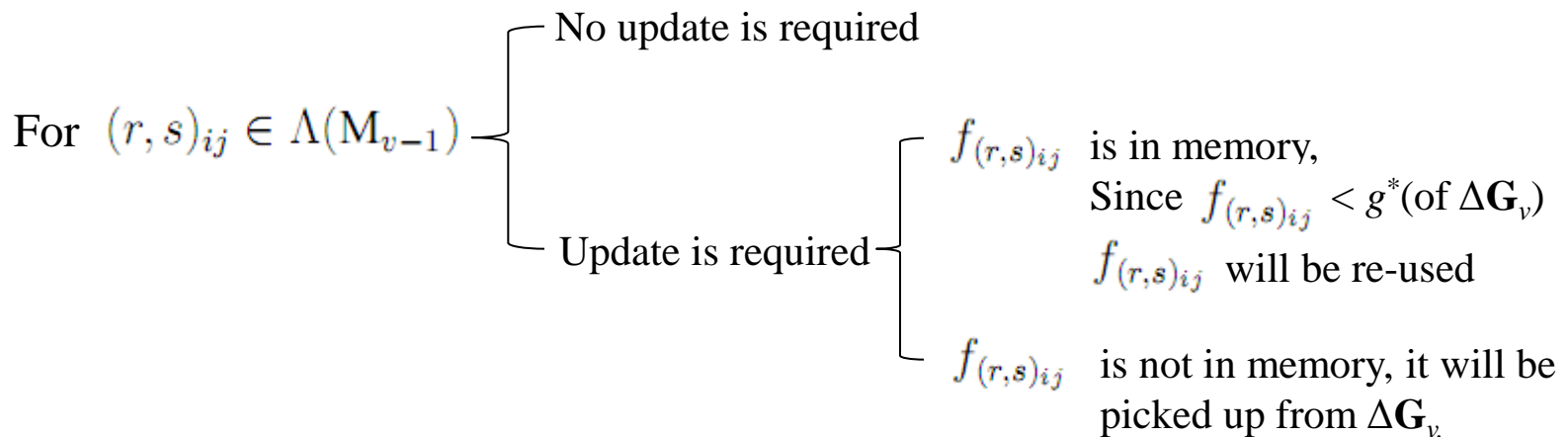
III. Validity Analysis



- *Lemma 1:* For all the polynomials $f_{(r,s)ij}$ with $(r,s)ij \in \Lambda(M_{v-1})$, we have

$$\text{lod}(f_{(r,s)ij}) < \text{lod}(y^{l_{v-1}+1}) \implies (f_{(r,s)ij} < y^{l_{v-1}+1})$$

- Expansion I: update $\Delta\mathbf{G}_v$ to $\Delta\tilde{\mathbf{G}}_v$ w.r.t. constraint of $\Lambda(M_{v-1})$



- The identity of the existing $f_{(r,s)ij}$ is left unchanged. Consequently, the solution of $\tilde{\mathbf{G}}_{v-1}$ remains intact.

- Therefore, $\mathbf{G}_v = \tilde{\mathbf{G}}_{v-1} \cup \Delta\tilde{\mathbf{G}}_v$

IV. Complexity Analysis



- Average decoding complexity – average number of finite field arithmetic operation for decoding one codeword frame;
- \mathcal{P}_{l_v} --- the probability of the decoder is performing a successful decoding with an OLS of l_v ;
- \mathcal{O}_{l_v} --- the decoding complexity with an OLS of l_v ;

- The average decoding complexity is:

$$\mathcal{O}_{\text{PASD}} = \underbrace{\sum_{v=1}^T \mathcal{P}_{l_v} \mathcal{O}_{l_v}}_{\text{Decoding succ.}} + \underbrace{(1 - \sum_{v=1}^T \mathcal{P}_{l_v}) \mathcal{O}_{l_T}}_{\text{Decoding fail.}}$$

- $\mathcal{O}_{\text{PASD}}$ is now channel dependent!

IV. Complexity Analysis



- *Theorem 2:* The decoding complexity of running the PASD algorithm with an OLS of l_v is:

$$\mathcal{O}_{l_v} = O(\mathcal{C}^2(\mathbf{M}_v)(l_v + 1)).$$

where $\mathcal{C}(\mathbf{M}) = |\Lambda(\mathbf{M})| = \frac{1}{2} \sum_{i=0}^{q-1} \sum_{j=0}^{n-1} m_{ij}(m_{ij} + 1)$

- \mathcal{O}_{l_v} consists of $\mathcal{O}_{l_v}^{\text{int}}$ and $\mathcal{O}_{l_v}^{\text{fac}}$

$$\mathcal{O}_{l_v}^{\text{int}} = O(\mathcal{C}(\mathbf{M}_v)(\mathcal{C}(\mathbf{M}_v) + 1)(l_v + 1)) \cong O(\mathcal{C}^2(\mathbf{M}_v)(l_v + 1))$$

$$\mathcal{O}_{l_v}^{\text{fac}} = O\left(k \sum_{\eta=1}^v (\mathcal{C}(\mathbf{M}_\eta) + 1)l_\eta\right) < O(kv(\mathcal{C}(\mathbf{M}_v) + 1)l_v)$$

since $kv \ll \mathcal{C}(\mathbf{M}_v)$

$$\mathcal{O}_{l_v}^{\text{fac}} \ll \mathcal{O}_{l_v}^{\text{int}}$$

$$\mathcal{O}_{l_v} \cong \mathcal{O}_{l_v}^{\text{int}}$$

IV. Complexity Analysis



- *Corollary 3:* When l_v is sufficiently large, the decoding complexity \mathcal{O}_{l_v} becomes:

$$\mathcal{O}_{l_v} = O\left(\frac{(k-1)^2}{4}(l_v^5 + l_v^4)\right)$$

- The decoding complexity increases exponentially with the OLS (dominant exponential factor of 5);
- The decoding complexity is quadratic in the dimension of the code k . The decoding complexity will be smaller for a low rate code.

IV. Complexity Analysis



- *Theorem 4:* The probability of the PASD algorithm performing a successful decoding with an OLS of l_v is given as:

$$\mathcal{P}_{l_v} = \Pr[S_{M_v}(\bar{c}) > \Delta_{1,k-1}(C(M_v)) \text{ and } S_{M_{v-1}}(\bar{c}) \leq \Delta_{1,k-1}(C(M_{v-1}))]$$

- The probability that the algorithm terminates at iteration v .

$$\text{Iter. 1: } \mathbf{S}_{M_1}(\mathbf{c}) \leq \Delta(C(M_1))$$

$$\text{Iter. 2: } \mathbf{S}_{M_2}(\mathbf{c}) \leq \Delta(C(M_2))$$

⋮

⋮

$$\text{Iter. } v-1: \mathbf{S}_{M_{v-1}}(\mathbf{c}) \leq \Delta(C(M_{v-1}))$$

$$\text{Iter. } v: \mathbf{S}_{M_v}(\mathbf{c}) > \Delta(C(M_v)) \longrightarrow \text{Exit!}$$

~~⋮~~

IV. Complexity Analysis



- Determining a closed form expression of \mathcal{P}_{l_v} turns out to be *hard*;
- Motivation of the analysis: link \mathcal{P}_{l_v} with Π (\sim channel SNR);

- Define $\mathcal{L} = \left\{ l \mid l > \frac{2 + \frac{\sqrt{n\gamma}}{\sqrt{\sum_{i,j} \pi_{ij}^2}}}{2\gamma[1 - \sqrt{k-1}\Phi_{\Pi}(\bar{c})]} \right\}$, where $\gamma = \frac{k-1}{n}$ and $\Phi_{\Pi}(\bar{c}) = \frac{\sqrt{\sum_{i,j} \pi_{ij}^2}}{S_{\Pi}(\bar{c})}$,

with OLS greater than the above threshold, successful decoding can be guaranteed.

- We therefore interpret \mathcal{P}_{l_v} as:

$$\mathcal{P}_{l_v} \cong \Pr[l_v = \min \mathcal{L}].$$

IV. Complexity Analysis



- Study the possible quantization of the OLS threshold $\frac{2 + \frac{\sqrt{n\gamma}}{\sqrt{\sum_{i,j} \pi_{ij}^2}}}{2\gamma[1 - \sqrt{k-1}\Phi_{\Pi}(\bar{c})]}$

- AWGN channel, we vary the SNR ...

In case of a ‘bad’ channel (e.g., SNR $\rightarrow -\infty$): $\pi_{ij} \cong 1/q$ for all $\pi_{ij} \in \Pi$.

$$\sqrt{\sum_{i,j} \pi_{ij}^2} \cong \sqrt{\frac{n}{q}}, \quad S_{\Pi}(\bar{c}) \cong \frac{n}{q}, \quad \Phi_{\Pi}(\bar{c}) \cong \sqrt{\frac{q}{n}}$$

In case of a ‘good’ channel (e.g., SNR $\rightarrow +\infty$): $\pi_{ij} \cong 1$ if $i = i_j$, and $\pi_{ij} \cong 0$ otherwise

$$\sqrt{\sum_{i,j} \pi_{ij}^2} \cong \sqrt{n}, \quad S_{\Pi}(\bar{c}) \cong n, \quad \Phi_{\Pi}(\bar{c}) \cong \frac{1}{\sqrt{n}}$$

- By refining $\frac{1}{\sqrt{n}} \leq \Phi_{\Pi}(\bar{c}) < \frac{1}{\sqrt{k-1}}$,

we can see the OLS threshold is a decreasing function of SNR.

- Recall $\mathcal{P}_{l_v} \cong \Pr[l_v = \min \mathcal{L}]$, $\mathcal{L} = \left\{ l \mid l > \frac{2 + \frac{\sqrt{n\gamma}}{\sqrt{\sum_{i,j} \pi_{ij}^2}}}{2\gamma[1 - \sqrt{k-1}\Phi_{\Pi}(\bar{c})]} \right\}$

\mathcal{P}_{l_v} will be in favor of smaller l_v values by increasing SNR!

IV. Complexity Analysis



- Recall the average decoding complexity definition:

$$\mathcal{O}_{\text{PASD}} = \sum_{v=1}^T \mathcal{P}_{l_v} \mathcal{O}_{l_v} + (1 - \sum_{v=1}^T \mathcal{P}_{l_v}) \mathcal{O}_{l_T}.$$

In a sufficiently ‘good’ channel:

$$\mathcal{P}_{l_1} \rightarrow 1 \text{ and } \mathcal{O}_{\text{PASD}} \cong \mathcal{O}_{l_1}$$

In a ‘bad’ enough channel:

$$\mathcal{P}_{l_1} \mathcal{P}_{l_2} \dots \rightarrow 0 \text{ and } \mathcal{O}_{\text{PASD}} \cong \mathcal{O}_{l_T}$$

IV. Simulation Statistics



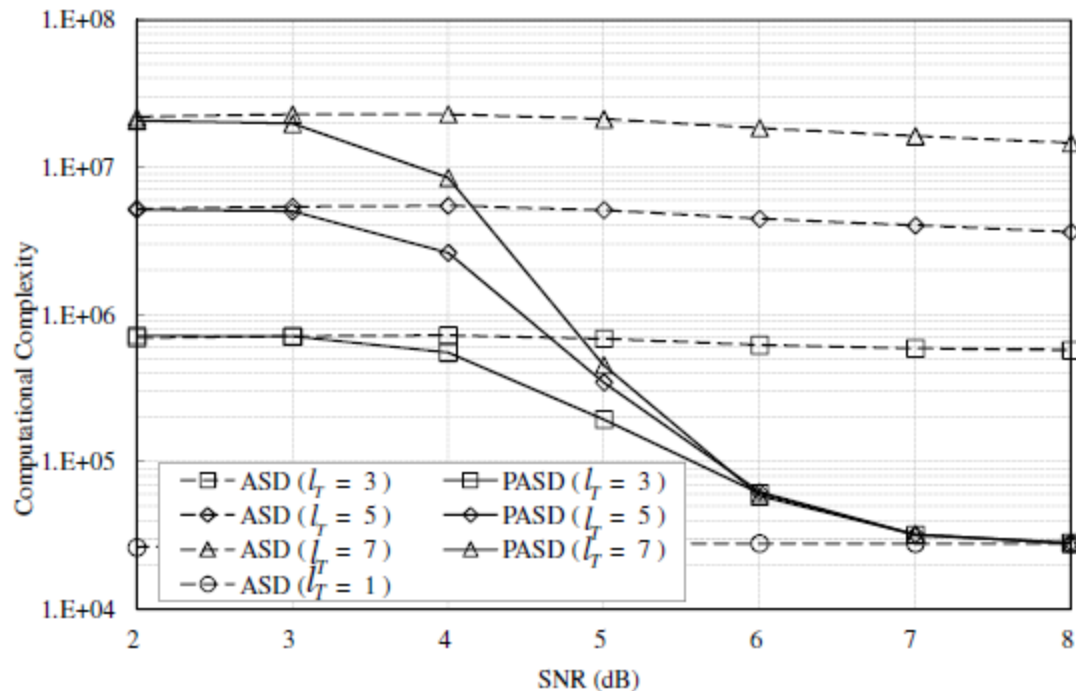
- $\mathcal{P}_{l_v} \sim$ SNR for the (63, 47) RS code with $l_1 = 1$, $l' = 1$ and $l_T = 5$
- AWGN channel

SNR (dB) \ \mathcal{P}_{l_v} (%)	2	3	4	5	6	7	8
\mathcal{P}_{l_1}	0.00	0.00	3.71	32.54	78.55	97.13	99.87
\mathcal{P}_{l_2}	0.00	1.51	28.47	56.30	21.25	2.87	0.13
\mathcal{P}_{l_3}	0.00	1.61	15.92	6.59	0.17	0.00	0.00
\mathcal{P}_{l_4}	0.00	1.23	9.68	1.95	0.02	0.00	0.00
\mathcal{P}_{l_5}	0.00	1.22	5.54	0.77	0.00	0.00	0.00

IV. Simulation Results



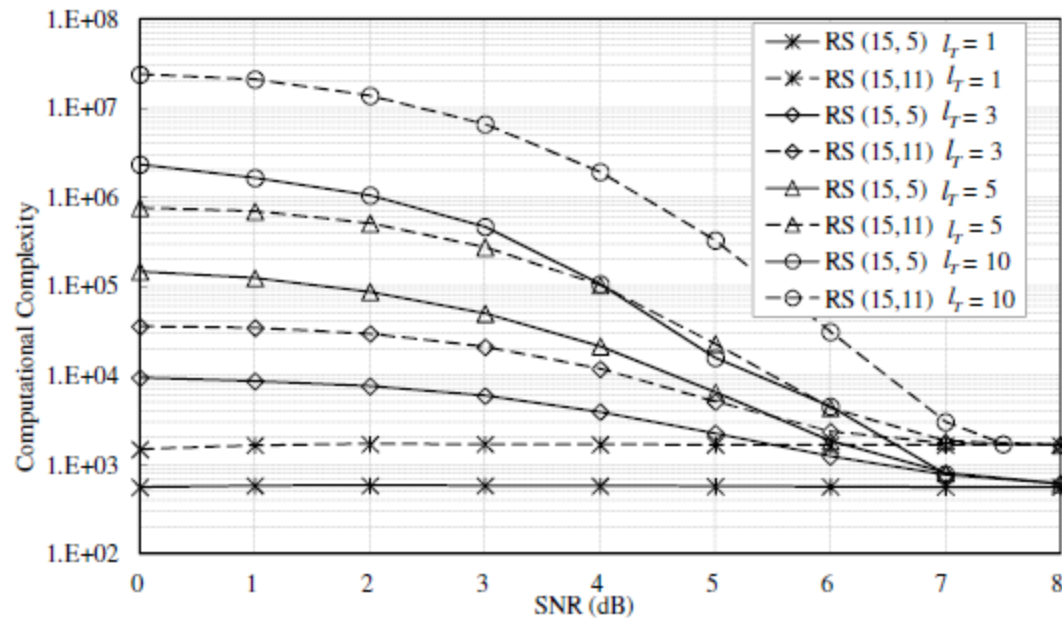
- $\mathcal{O}_{\text{PASD}} \sim \text{SNR}$ for the (63, 47) RS code with $l_T = 3, 5, 7$.
- AWGN channel



IV. Simulation Results



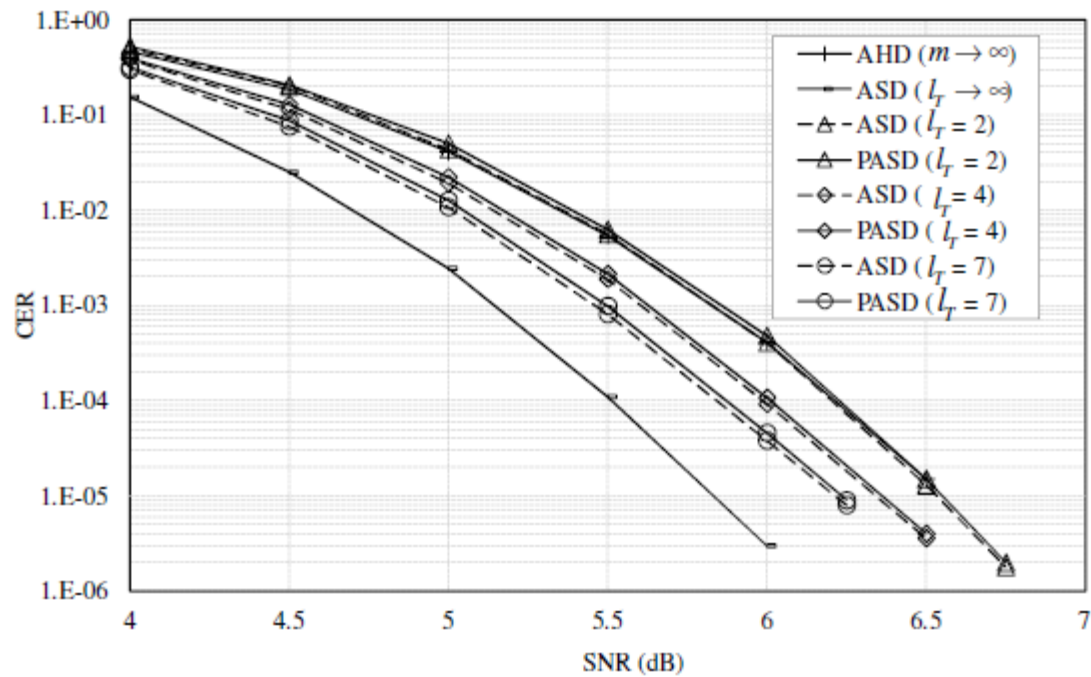
- $O_{\text{PASD}} \sim \text{SNR}$ for the (15, 11) and (15, 5) RS codes with $l_T = 1, 3, 5, 10$
- AWGN channel



V. Error-Correction Performance



- PASD ~ ASD with same decoding parameter l_T ;
- For PASD, decoding output is validated by CRC code;
- For ASD, decoding output is validated by the ML criterion;
- For the (63, 47) RS code, over AWGN channel:



VI. Conclusions



- A progressive algebraic soft-decision decoding approach;
- Two key steps of PASD: progressive reliability transform & progressive interpolation;
- Enables the complexity dominant interpolation process to be performed in an iterative manner, and the incremental computation between iterations is possible;
- The average decoding complexity of the PASD algorithm is channel dependent and hence it has been optimized according to the needs;
- Error-correction performance is also preserved.
- A larger system memory is required.

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Project: Research on adaptable list decoding system;
ID: 10451027501005078; From 2010. 10 to 2012. 10.